# 3D FINITE-ELEMENT SIMULATION OF CONTROLLED-SOURCE ELECTROMAGNETIC PROBLEMS USING EDGE AND NODAL INTERPOLATION FUNCTIONS Seyedmasoud Ansari and Colin Farquharson

# SUMMARY

#### We deal with the three dimensional finite-element solution of the CSEM problems in a conductive medium. 3D unstructured tetrahedral meshes are used to discretize the computational domain. Inorder to separate the galvanic and inductuve natures of the EM field, an ${f A}$ - $\phi$ decomposition of the electric field is done. After obtaining two partial differential equations, the finite-element approximation of the system of equations is done using edge and nodal element basis functions.

# METHODOLOGY

Assuming a time dependence of  $e^{i\omega t}\,$  , Faraday's law and Ampere's laws of induction are written as:

$$\nabla \times \mathbf{E} + i\omega\mu\mathbf{H} = 0$$

(1)

$$abla imes \mathbf{H} - \tilde{\sigma} \mathbf{E} = \mathbf{J}^s$$

 $\mu$  = Magnetic Permeability

 $\tilde{\sigma}$  = Electrical Conductivity

 $\mathbf{J}^{s}$  = Source Current Density

Upon combining equations (1) and (2), a system of equations is obtained

$$\nabla \times \nabla \times \mathbf{E} + i\omega\mu(\tilde{\sigma}\mathbf{E} + \mathbf{J}^s) = 0 \qquad (3)$$

$$abla \cdot \mathbf{J} = -
abla \cdot \mathbf{J}^s$$

(4)

Decomposition of the electric field:

$$\mathbf{E} = -i\omega\mathbf{A} - \nabla\phi$$

 $\mathbf{A}$  = Magnetic Vector Potential

 $\phi$  = Electric Scalar Potential

Decomposed Equations in the form of Partial Differential Equations:

$$\nabla \times \nabla \times \mathbf{A} + i\omega\mu\tilde{\sigma}\mathbf{A} + \mu\tilde{\sigma}\nabla\phi = \mu\mathbf{J}^s$$

 $\left|-i\omega\nabla\cdot\left(\tilde{\sigma}\mathbf{A}\right)-\nabla\cdot\left(\tilde{\sigma}\nabla\phi\right)=-\nabla\cdot\mathbf{J}^{s}\right|$ (6)

# DISCRETIZATION USING UNSTRUCTURED TETRAHEDRAL ELEMENTS

"Blocks2mesh", a code written by Peter Lelièvre (plelievre@mun.ca), is utilized to create a three dimensional grid of unstructured tetrahedrons. This code makes a block poly file that then get fed into *Tetgen*. Global numbers are assigned to each element and their corresponding nodes within the entire domain.

Because we mostly need the edge numbering scheme in our approach, a Fortran code is developed to convert the node numbering scheme into the edge one. Also, some auxiliary subroutines are written to figure out if any edges and nodes are shared by other neighbour cells. Moreover, edge-to-node, edge-to-element, element-to-edge, and other similar connectivity arrays have been coded.

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Figure 1. The finite-element mesh mesh for a homogenous Earth model depicted using Kitware Paraview 3.6.2. This mesh consists of 1233 nodes, 4432 cells and 10753 edges.





Tetrahedrons are more suitable to model the curves of the complex-shaped volumes and irregular geometries, for instance, jagged topography features, faults and dipping layers. The finite-element method uses nodal and edge basis functions to determine the approximated solution for the electric field within each element.



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#### THE FINITE-ELEMENT APPROACH

Doing the finite-element solution of the system of equations requires the approximation of the vector and scalar potentials as follows:

$$\tilde{\mathbf{A}} = \sum_{j=1}^{N_{edges}} \tilde{A}_j \mathbf{N_j}$$
(7) 
$$\mathbf{N_j} = \text{Edge-element vector interpolation function}$$
$$\tilde{\phi} = \sum_{k=1}^{N_{nodes}} \tilde{\phi_{k_1}} N_{k_1}$$
(8) 
$$N_{k_1} = \text{Nodal-element scalar interpolation function}$$

**The Galerkin Method**: Seeks the solution by wighting the vector and scalar residuals of the partial differential equations

#### **1-Vector Residual**

$$\mathbf{R} = \int_{\Omega} \mathbf{N}_{i} \cdot \mathbf{r} \ d\Omega = 0 \qquad (9) \qquad \mathbf{r} = \nabla \times \nabla \times \mathbf{A} + i\omega\mu\tilde{\sigma}\mathbf{A} + \mu\tilde{\sigma}\nabla\phi - \mu\mathbf{J}^{s}$$

$$\sum_{j=1}^{N_{edges}} \tilde{A}_{j} \int_{\Omega} (\nabla \times \mathbf{N}_{i}) \cdot (\nabla \times \mathbf{N}_{j}) \ d\Omega + i\omega\mu \sum_{j=1}^{N_{edges}} \tilde{A}_{j} \int_{\Omega} \tilde{\sigma}\mathbf{N}_{i} \cdot \mathbf{N}_{j} \ d\Omega$$

$$+\mu \sum_{k=1}^{N_{nodes}} \phi_{k} \int_{\Omega} \tilde{\sigma}\mathbf{N}_{i} \cdot \nabla N_{k_{1}} \ d\Omega = \mu \sum_{j=1}^{N_{edges}} \int_{\Omega} \mathbf{N}_{i} \cdot \mathbf{J}_{s} \ d\Omega \qquad (10)$$

#### 2- Scalar Residual

$$R_{scalar} = \int_{\Omega} N_{k_2} r \ d\Omega = 0 \qquad (11) \qquad r = i\omega \nabla \cdot (\tilde{\sigma} \mathbf{A}) + \nabla \cdot (\tilde{\sigma} \nabla \phi) - \nabla \cdot \mathbf{J}^s$$

$$i\omega \tilde{A}_j \sum_{j=1}^{N_{edges}} \int_{S} N_{k_2} \hat{n} \cdot (\tilde{\sigma} \mathbf{N}_j) \ dS - i\omega \tilde{A}_j \sum_{j=1}^{N_{edges}} \int_{\Omega} \nabla N_{k_2} \cdot (\tilde{\sigma} \mathbf{N}_j) \ d\Omega \qquad (12)$$

$$+ \tilde{\phi}_{k_1} \sum_{k_1=1}^{N_{nodes}} \int_{S} N_{k_2} \hat{n} \cdot (\tilde{\sigma} \nabla N_{k_1}) \ dS - \tilde{\phi}_{k_1} \sum_{k_1=1}^{N_{nodes}} \int_{\Omega} \nabla N_{k_2} \cdot (\tilde{\sigma} \nabla N_{k_1}) \ d\Omega$$

$$= \int_{\Omega} N_{k_2} \nabla \cdot \mathbf{J}_s \ d\Omega$$

#### NODAL ELEMENTS

Within each linear tetrahedral element, the unknown scalar function can be written as

$$\phi^{e}(x, y, z) = a^{e} + b^{e}x + c^{e}y + d^{e}z$$
 (1)

Coefficients  $a^e$ ,  $b^e$ ,  $c^e$ , and  $d^e$  can be determined by enforcing the equation (13) at four corners of the tetrahedron. By substituiting these coefficients into equation (13) we obtain:

$$\phi^{e}(x, y, z) = \sum_{k=1}^{4} N_{k}^{e}(x, y, z)\phi_{j}^{e}$$
(14)

 $N_k^e$  = Scalar basis function

$$V_k^e(x, y, z) = \frac{1}{6V^e} (a_k^e + b_k^e x + c_k^e y + d_k^e z)$$
(15)

 $V^e$  = The volume of the tetrahedron

As the main property, the value of the interpolation function  $\left[N_k(x_i,y_i,z_i)
ight]$ is specified to be 1 if i = k and zero if  $i \neq k$ .



Figure 2. An illustration of the scalar basis function related to the node i. Scalar N decreases linearly in different directions away from its initial location and vanishes at the location of nodes j, l, and k. In contrast to the horizontal component of the N<sub>iz</sub>, which is depicted by  $N_{iz}^{hor}$ , its vertical component, N  $\frac{vec}{iz}$ , is discontinous across the edge "il".

### EDGE-ELEMENTS

Nodal basis functions guarantee only the continuity of the tangential component of the electric field across the interface while the vertical component of the electric field is not necessarily continuous. This causes the normal current density to be no longer continous across the interface. Also, Because the electric field is allowed to vary in all three orthogonal directions, the term  $\nabla \cdot (\sigma \mathbf{E})$  is not consistently zero within cells of uniform conductivity. To counteract these problems edge-element basis functions are used. Constrained to be along the edge, an edge-element combines three components of the approximated field into one collection. This guarantees the continuity of the tangential component, while allowing the normal component of the electric field to freely jump across the interelement boundaries.

The Whitney 1-form is used as the vector edge-element function. By using this form the basis function, for example, along edge 1 (see Fig. 3) is given by:

$$\mathbf{P}_{12} = L_1^e \nabla L_2^e - L_2^e \nabla L_1^e \tag{16}$$

 $L_k^e = N_k$  ( k = 1, 2, 3, and 4)

Where

#### Advantage 1

**P**<sub>12</sub> has no tangentila component along other edges exept edge 1, thus, forcing the tangential component of the field to be continuous between adjacent cells sharing this edge.

In general, the edge function is given by:

$$\mathbf{N_{i}^{e}} = l_{i}^{e} (L_{i1}^{e} \nabla L_{i2}^{e} - L_{i2}^{e} \nabla L_{i1}^{e})$$
(17)

Advantage 2:

$$abla \cdot \mathbf{N_i} = 0$$

indicates that the field is divergence-free in a source-free region



Figure 3. Six edges are depicted for a tetrahedral element. Edge 1 is picked as an example which starts from node 1 and ends at node 2. It has no tangential component on the blue surface  $S_2$ 

## CALCULATION OF RESIDULAS USING ELEMENTAL FUNCTIONS

$$\mathbf{T}_{ij} = \sum_{j=1}^{N_{edges}} \int_{\Omega} \left( \nabla \times \mathbf{N}_{\mathbf{i}} \right) \cdot \left( \nabla \times \mathbf{N}_{\mathbf{j}} \right) \, d\Omega \qquad \qquad \mathbf{i} = 1, \dots \, \text{Nedges}$$
(18)

$$\mathbf{T}_{ij} = \sum_{j=1}^{N_{edges}} \frac{4 \ l_i \ l_j}{(6)^4 (V^e)^3} \left[ (b_{i1}c_{i2} - c_{i1}b_{i2})(b_{j1}c_{j2} - c_{j1}b_{j2}) \right]$$
(19)

$$+(d_{i1}b_{i2}-b_{i1}d_{i2})(d_{j1}b_{j2}-b_{j1}d_{j2}) + (c_{i1}d_{i2}-d_{i1}c_{i2})(c_{j1}d_{j2}-d_{j1}c_{j2})]$$

To easily calculate the second edge - edge interaction integral we changed the coordinate system from cartesian to **simplex or normalized** system

3 - Node - Node Interactions

$$\mathbf{Z_{k2k1}} = \sum_{k_1=1}^{N_{nodes}} \int_{\Omega} \nabla N_{k_2} \cdot \tilde{\sigma} \nabla N_{k_1} \, d\Omega \tag{24}$$

$$\mathbf{Z_{k2k1}} = \tilde{\sigma}(\frac{1}{6V^e})[b_{k2}b_{k1} + c_{k2}c_{k1} + d_{k2}d_{k1}]$$
(25)

#### ELECTROMAGNETIC SOURCE

A line source of electric current is chosen to be as a delta function of finite length :

$$\mathbf{J} = \delta(y') \ \delta(z') \ Box(x')j_s \ \mathbf{\hat{x}'}$$
(26)

where  $\dot{J}_{s}$  is a scalar arbitrary function and Box (x') is the boxcar function; it indicates that the source current is zero over the entire domain except for a specific interval, I, on which it is given a constant value.

(27)

#### Source - Edge interaction :

$$\mathbf{S} = \int_{\Omega} \mathbf{N}_{\mathbf{i}} \cdot \mathbf{J} \ d\Omega$$

The secondary system, (x', y', z') is chosen in a way that the source line has the simplest equation. Because the tetrahedral elements are set up in the cartesian coordinate system a transformation has to be done in order to easily calculate the source term integrals.

$$\mathbf{S} = \mu \sum_{i=1}^{N_{edges}} \left(\frac{1}{6V^e}\right)^2 (a_{i1}b_{i2} - a_{i2}b_{i1})j_s l \qquad (28)$$

SS =

$$\mathbf{SS} = \int_{\Omega} N_{k_2} \nabla \cdot \mathbf{J}_{\mathbf{s}} \, d\Omega \qquad (2$$

$$\mathbf{SS} = \int_{\Omega} N_{k_2} \nabla \cdot \mathbf{J_s} \ d\Omega$$
$$\int_{\Omega} N_{k_2} \nabla \cdot \mathbf{J_s} \ d\Omega = \sum_{k=1}^{N_{nodes}} b_k \ j_s(\frac{1}{6V^e}) \ l \qquad (30)$$



cartesian and target coordinate system

## NUMERICAL SOLUTION TO THE SYSTEM OF EQUATIONS

Plugging equations (19), (21), (23), (25), (28), and (30) into both equations (10) and (12) gives us a system of equations in the form of :

$$L.u = F' \tag{31}$$

Separating the potentials into their real and imaginary parts, we end up with the following equations.

$$\begin{pmatrix} \mathbf{T}_{\mathbf{ij}} & -\omega\mu\mathbf{U}_{\mathbf{ij}} & \mu\mathbf{W}_{\mathbf{ik_1}} & 0 \\ \omega\mu\mathbf{U}_{\mathbf{ij}} & \mathbf{T}_{\mathbf{ij}} & 0 & \mu\mathbf{W}_{\mathbf{ik_1}} \\ 0 & \omega\mathbf{W}_{\mathbf{k_2j}} & -\mathbf{Z}_{\mathbf{k_2k_1}} & 0 \\ -\omega\mathbf{W}_{\mathbf{k_2j}} & 0 & 0 & -\mathbf{Z}_{\mathbf{k_2k_1}} \end{pmatrix} \begin{pmatrix} \mathbf{A}_{\mathbf{j}}^{\mathbf{r}} \\ \mathbf{A}_{\mathbf{j}}^{\mathbf{I}} \\ \phi_{k}^{r} \\ \phi_{k}^{I} \end{pmatrix} = \begin{pmatrix} \mathbf{S} \\ 0 \\ \mathbf{SS} \\ 0 \end{pmatrix}$$
(32)

#### **Boundary Conditions:**

Boundary conditions are applied on the truncation boundary of the domain. Subject to the Dirichlet boundary condition, if the source is considered far enough from the truncation boundary, the following equations will be satisfied.

$$\mathbf{A} \times \mathbf{A}|_{\partial_{\Omega}} = \mathbf{0} \qquad \phi|_{\partial_{\Omega}} = 0$$

To code up the system of equations a **Fortran** code is written. The sparse structure of the coefficient matrix and the non zero elements are shown by the figure 6.



Figure 6. The sparse structure of the coefficient matrix. It is designed for a mesh of 216 cells, 80 nodes and 366 edges. This matrix consists of four main blocks. The dominant block, (1:732, 1:732), is related to the edge-edge interactions (top left block). Also, the dimension of the matrix is 2(# of Edges + # of Nodes) = 892

The nonzero values of the sparse coefficent matrix on the left is stored in the CSR format for the sake of lower memory usage. The non-symmetric system of equations will be solved using **BICGSTAB** solver with an **LU** preconditioner.

# **CONCLUSIONS AND FUTURE PLANS**

The code is still under development. We hope to test it on simple Earth models. The performance of the discussed approach will be compared with a modification in which the Lorentz gauge condition is used. Also, the relative contributions to the electric field from the inductive and galvanic terms will be investigated in different situations.

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